



#### FEATURES

- Dual channel CW operation with up to 750 mA per channel
- Up to 1500 mA with both channels combined
- ♦ 2.8 V to 11 V power supply
- ♦ Operation with or without µController
- Individual enable input per channel
- ♦ Control loop accuracy better than 1%
- Internal programmable logarithmic monitor resistor
- Operating point setup with 10 bit logarithmic resolution
- ACC or APC mode individually configurable for each channel
- ♦ A/D converters for analog monitoring
- Serial programming interface (SPI or I<sup>2</sup>C compliant)
- Configuration content verification and validation
- Programmable laser overcurrent protection
- Optimized for N-type laser diodes
- ♦ Low drop linear regulator for 3.3 V
- Low current standby mode
- Temperature monitor
- ♦ Temperature range -40 ... 85 °C



#### APPLICATIONS

- Laser diode and LED modules
- ♦ CW N-/M-type laser diode drivers
- Embedded laser diode controllers
- Structured-light 3D illuminations
- Multiple laser diode control
- Optical amplification/pumping
- Safety related laser controllers

PACKAGES



#### QFN28 5 mm x 5 mm



05/24 / V5 / MaH-IF / ichaus/ht-dual-cw-laser-diode-driver

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#### DESCRIPTION

Dual CW laser diode driver iC-HT can operate two individual laser diodes with up to 750 mA laser current depending on the heat dissipation. Each channel can be enabled independently. The laser diode driver can be controlled by an external microcontroller (MCU mode) or operate stand alone with pin/resistor configuration (iC-WK mode). In MCU mode, both channels can be combined for driving up to 1500 mA.

Each channel can be operated individually either in automatic current control (ACC) or automatic power control (APC). All parameters including the internal reference voltages are set via serial communication ( $l^2C$  or SPI). A 10 bit resolution D/A converter with logarithmic characteristic is used for setting the operating point. This allows an operating point resolution better than 1%.

In APC control, the monitor diode photocurrent is used to track the optically emitted power of the laser diode. The feedback for the laser diode driver is the voltage of the photocurrent at a monitor resistor. An 8 bit internal programmable logarithmic monitor resistor (PLR) or an external monitor resistor can be selected for closing the control loop. The PLR ranges from 100  $\Omega$  to 500 k $\Omega$  with a step width less than 5%.

In ACC control, the laser diode current can be set directly. Two current ranges are selectable.

iC-HT allows disabling the laser channels when an overcurrent threshold has been exceeded. The over-

current threshold of each channel has 2 ranges and is programmable through an 8 bit linear D/A converter.

The temperature monitor measures the internal chip temperature. iC-HT disables the laser channels when overtemperature is detected.

A variety of voltages can be measured with a 10 bit A/D converter. The following voltages can be measured:

- V(LDKx)
- V(VDD)
- V(VB)
- V(MDAx)
- V(PLRx)

The DCO current output pin can control an external DC/DC converter. Controlling the DC/DC output voltage can optimize the power dissipation of the whole system e.g. to extend battery life.

iC-HT in standby mode has a very low current consumption (< 10  $\mu$ A) and does retain its configuration.

The device features for safe operation:

- Configuration verification
- · Tri-state configuration pins
- · Write protection in operating mode
- · Safe default/startup state

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#### PACKAGING INFORMATION QFN28 5 mm x 5 mm to JEDEC

PIN CONFIGURATION QFN28 5 mm x 5 mm (topview)



PIN FUNCTIONS Function No. Name 1 LDK1 Laser Diode Cathode for channel 1 2 LDK1 Laser Diode Cathode for channel 1 Analog Ground for channel 1 3 AGND1 Integration Capacitor for channel 1 4 CI1 5 CIL1 Integration Capacitor for channel 1, low side Monitor Diode Anode for channel 1 6 MDA1 7 MRL1 Monitor Resistor for channel 1, low side 8 EMC Enable Microcontroller input 9 SCLK/SCL SPI Clock / I<sup>2</sup>C Clock 10 MISO/SDA SPI Master In Slave OUT / I<sup>2</sup>C Data SPI Master Out Slave In / I<sup>2</sup>C Ad-11 MOSI/A0 dress Bit 0 12 NCS/A1 Chip Select, active low / I<sup>2</sup>C Address bit 1 Enable Channel 1 input 13 EC1 14 EC2 Enable Channel 2 input 15 MRL2 Monitor Resistor for channel 2, low side 16 MDA2 Monitor Diode Anode for channel 2 Integration Capacitor for channel 2, 17 CIL2 low side 18 CI2 Integration Capacitor for channel 2 19 AGND2 Analog Ground for channel 2 20 LDK2 Laser Diode Cathode for channel 2 21 LDK2 Laser Diode Cathode for channel 2 22 GND Ground 23 DCO **Digital Current Output** I<sup>2</sup>C or SPI selection input / Refer-24 INS/WKR ence voltage selection in iC-WK mode 25 VDD 3.3 V output supply 26 VB Power supply Error output, active low 27 NCHK 28 NSTBY Standby input, active low TP Thermal Pad (GND)

The *Thermal Pad* is to be connected to a *Ground Plane* (GND, AGND1...2) on the PCB. Only pin 1 marking on top or bottom defines the package orientation (**©** HT label and coding is subject to change).

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# iC-HT DUAL CW LASER DIODE DRIVER Rev C1, Page 4/46 PACKAGE DIMENSIONS QFN28-5x5 All dimensions given in mm. This package falls within JEDEC MO-220-VHHD-1. RECOMMENDED PCB-FOOTPRINT 4.70 3.15 R0.15 UUUUUUUUUU SIDE 0.90 3.15 4.70 <u> = 0.0.0.0.0.0</u> E Ш <del>┨</del>╼┣ 0.90 0.50 0.30 TOP BOTTOM 5 3.15 UUUUUU0 3.15 ഹ 7 $\subset$ D C $\square$ $\square$ Π $\cap$ 0.55 0.50 0.25 drb\_qfn28-2\_pack\_1, 10:1

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#### ABSOLUTE MAXIMUM RATINGS

These ratings do not imply permissible operating conditions; functional operation is not guaranteed. Exceeding these ratings may damage the device.

Item	Symbol	Parameter	Conditions			Unit
No.	-			Min.	Max.	
G001	VB	Voltage at VB		-0.3	11	V
G002	I(VB)	Current in VB		-20	50	mA
G003	VDD	Voltage at VDD		-0.3	5.5	V
G004	I(VDD)	Current in VDD		-20	1	mA
G005	V()	Voltage at Cl1, Cl2, EC1, EC2, MDA1, MDA2, EMC, SCLK/SCL, MISO/SDA, MOSI/A0, NCS/A1, DCO, INS/WKR, NCHK		-0.3	5.5	V
G006	1()	Current in Cl1, Cl2, EC1, EC2, MDA1, MDA2, EMC, SCLK/SCL, MISO/SDA, MOSI/A0, NCS/A1, DCO, INS/WKR, NCHK, NSTBY, CIL1, CIL2, MRL1, MRL2		-20	20	mA
G007	I(LDK)	Current in LDK1, LDK2	DC current	-20	900	mA
G008	V()	Voltage at LDK1, LDK2, NSTBY		-0.3	11	V
G009	V()	Voltage at AGND1, AGND2		-0.3	1	V
G010	I(AGND)	Current in AGND1, AGND2	DC current	-900	1	mA
G011	Vd()	ESD Susceptibility at all pins	HBM 100 pF discharged through $1.5 \text{ k}\Omega$		2	kV
G012	Tj	Operating Junction Temperature		-40	125	°C
G013	Ts	Storage Temperature Range		-40	150	°C

#### THERMAL DATA

Operat	Operating Conditions: VB = 2.8 11 V (referenced to GND)						
ltem	n Symbol Parameter Conditions					Unit	
No.				Min.	Тур.	Max.	
T01	Та	Operating Ambient Temperature Range		-40		85	°C
T02	Rthja	Thermal Resistance Chip/Ambient	Mounted on PCB		25		K/W
T03	RthjTP	Thermal Resistance Chip/Thermal Pad			4		K/W

All voltages are referenced to ground unless otherwise stated. All currents flowing into the device pins are positive; all currents flowing out of the device pins are negative.

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#### ELECTRICAL CHARACTERISTICS

Item Symbol Parameter Conditions							
No.	-			Min.	Тур.	Max.	
Functi		parameters beyond the operating ndividual application using FMEA n	conditions (with reference to independent von nethods.	ltage suppli	es, for ir	istance)a	re to b
001	VB	Permissible Supply Voltage	Referenced to GND	2.8		11	V
002	I(VB)	Standby Current at VB	V(NSTBY) < 0.4 V	_		10	μA
003	I(VB)	Supply Current at VB	No load, EC1, EC2, NSTBY = hi			5	mA
004	V(VB)on	Turn-on threshold	Increasing VB	1.6		2.7	v
005	V(VB)off	Turn-off threshold	Decreasing VB	1.45		2.6	V
006	V(VB)Hys	Power-on hysteresis		20		250	mV
007	V(VDD)on	Turn-on threshold	Increasing VDD	1.6		2.4	V
008	V(VDD)off	Turn-off threshold	Decreasing VDD	1.45		2.3	V
009	V(VDD)Hys	Power-on hysteresis		20		250	mV
010		RAM memory reset during Stand-By	NSTBY = lo	0.85		1.4	V
011	Rgnd()	Resistor to GND at MRL1, MRL2				20	Ω
012	Ragnd()	Resistor to AGNDx at CILx				20	Ω
013	Vc()lo	Clamp Voltage Io at VB, VDD, NCHK, EMC, NCS/A1, MISO/SDA, MOSI/A0, SCLK/SCL, INS/WKR, NSTBY, EC1, EC2, DCO, LDK1, LDK2, CI1, CIL1, CI2, CIL2, AGND1, AGND2, MDA1, MDA2	l() = -10 mA	-1.6		-0.3	V
Laser	Driver LDK	c, Cix, MDAx					
101	1	Saturation Voltage lo at LDK	I(LDKx) = 750 mA			0.7	V
102	V(LDKSAT)	LDKx saturation detection threshold	RLDKSx = 00 RLDKSx = 01 RLDKSx = 10 RLDKSx = 11	0.35 0.55 0.85 1.05	0.5 0.7 1 1.2	0.65 0.85 1.15 1.35	V V V V
103	Idc(LDK)	Permissible DC Current at LDKx				750	mA
104	lleak(LDK)	LDKx leakage current	V(LDKx) = 11 V			10	μA
105	C(CI)	Possible capacitor at CI1, CI2	ECIE = 0, EMC = hi	0			μF
106	I(CI)	Charge Current at CI1, CI2	V(CI) = 0 V, EC1, EC2 = hi, ECIEx = 1 COMP = 111	-220		-30	μA
107	I(LDK)max	Laser overcurrent shutdown threshold	V(LDKx) = 0.7 V VB - 1.5 V ILIMx(7:0) = 0x00, RACCx = 0 ILIMx(7:0) = 0xFF, RACCx = 0 ILIMx(7:0) = 0x00, RACCx = 1 ILIMx(7:0) = 0xFF, RACCx = 1	0 750 0 80		25 1400 3.2 175	mA mA mA mA
108	⊿ I(LDK)	Shutdown threshold resolution	RACCx = 0 RACCx = 1	2.8 0.3	4 0.5	5.2 0.8	mA mA
109	tovc	Time to overcurrent shutdown	Laser current decreased 10%	1		5	μs
110	V(MDA)	Voltage at MDA1, MDA2	Closed control loop EC1, EC2 = hi EMC = lo, INS/WKR = lo EMC = lo, INS/WKR = hi	225 455	250 500	275 545	mV mV
111	Ten	Time to laser enabled	NSTBY lo $\rightarrow$ hi, no load at VDD, V(VDD) 0 to 90%, CVDD = 1 µF, EMC = lo			1.3	ms
112	Tci	Time to light	NSTBY = hi, ECIE = 0, COMP = 010, light off to 80% target value			300	μs
	Tcio	Time to target light	Light from 80% to 99% target value			4700	μs

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#### ELECTRICAL CHARACTERISTICS

ltem No.	Symbol	Parameter	Conditions	Min.	Тур.	Max.	Unit
114	ldc(LDK)	LDKx ACC mode current	EC1, EC2, EMC = hi, EACCx = 1, V(LDKx) = 0.7 V VB - 1.5 V REFx(9:0) = 0x000, RACCx = 0 REFx(9:0) = 0x30F, RACCx = 0 REFx(9:0) = 0x30F, RACCx = 1 REFx(9:0) = 0x30F, RACCx = 1	50 650 5 70	82 750 10 113	120 1400 15 160	mA mA mA mA
115	Tk	Temperature coefficient ACC mode		-1500	-500	0	ppm/ł
Progr	ammable Re	esistor					
201	Rmda	Resistor at MDAx pin	RMDx(7:0) = 0xF0 0xFF, DISPx = 0 RMDx(7:0) = 0x00 0x0F, DISPx = 0	350 0.154	500 0.220	650 0.286	kΩ kΩ
202	Tk	Temperature coefficient		-1500	-500	0	ppm/l
203	ΔR	Resistor increment	$\Delta R = \frac{R(n+1) - R(n)}{R(n)}$	2	3.3	5	%
204	lleak(MDA)	MDAx leakage current	DISPx = 1	-1		1	μA
D/A C	onverter		J				
301	R(DAC)	D/A Converter Resolution				10	bit
302	ΔV	Percentual voltage increments	$\Delta V = \frac{V(n+1) - V(n)}{V(n)}$	0.1	0.235	0.5	%
303	V(DAC)	D/A Converter	REFx(9:0) = 0x000 lowest value REFx(9:0) = 0x3FF highest value	0.09 1.00	0.10 1.10	0.12 1.25	V V
Check	Output NC	нк	J				
401	Vs()lo	Saturation Voltage lo at NCHK	I(NCHK) = 1.0 mA			0.4	V
402	lsc()lo	Short Circuit Current lo at NCHK	V(NCHK) = 0.4 3.3 V	9		33	mA
403	llk()	Leakage Current at NCHK	NCHK = 1; V(NCHK) = 0 5.5 V	-1		1	μA
Series	Regulator	Output VDD			1		
501	V(VDD)	Regulated output voltage	VB = 3.7 8 V, I(VDD) = -10 0 mA NSTBY = hi	3		3.5	V
502	V(VB,VDD)	Voltage Drop between VB and VDD	VDD unregulated, I(VDD) = -10 0 mA NSTBY = hi		100	400	mV
503	C(VOUT)	Capacitor at VDD	Ri(C) < 1 Ω	1		3.3	μF
504	Tvdd	Settling time VDD	NSTBY Io $\rightarrow$ hi, no load at VDD, V(VDD) 0 to 90 % CVDD = 1 $\mu$ F			1	ms
Digita	l inputs	•					
601	Vt()hi	Input Threshold Voltage hi at NCS/A1, MISO/SDA, MOSI/A0, SCLK/SCL, NSTBY, EC1, EC2				2	V
602	Vt()lo	Input Threshold Voltage Io at NCS/A1, MISO/SDA, MOSI/A0, SCLK/SCL, NSTBY, EC1, EC2	VB > 3 V VB = 2.8 V	0.7 0.6			V V
603	Vt()hys	Hysteresis at NCS/A1, MISO/SDA, MOSI/A0, SCLK/SCL, NSTBY, EC1, EC2	Vt()hys = Vt()hi - Vt()lo	100			mV
604	lpd()	Pull-Down Current at MOSI/A0, EC1, EC2	V() = 0.4 V VDD	1		50	μA
605	lpd()	Pull-Down Current at NSTBY	V() = 0.4 V VB	1	İ	50	μA
606	Rpu()	Pull-Up Resistor at SCLK/SCL, NCS/A1		80	150	260	kΩ
607	Rpu()	Pull-Up Resistor at MISO/SDA	EMC = hi, INS/WKR = lo EMC = hi, INS/WKR = hi	8 53	20 100	50 174	kΩ kΩ
608	Er()	Safe enable threshold voltage at EMC, INS/WKR	Rising Falling	52 30	54 32	56 34	% VD % VD
609	Voc()	Open Circuit Voltage at EMC, INS/WKR		39	41	43	% VD
610	Ri()	Internal Resistance at EMC, INS/WKR		170	250	330	kΩ

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#### ELECTRICAL CHARACTERISTICS

ltem No.	Symbol	Parameter	Conditions	Min.	Тур.	Max.	Unit
611	lsc()lo	Short Circuit current lo at MISO/SDA	INS/WKR = Io, V(MISO/SDA) = 5.5 V	-40		-4	mA
612	Vs()lo	Saturation Voltage lo at MISO/SDA	INS/WKR = Io, I(MISO/SDA) = 2 mA			0.4	V
A/D C	onverter						
701	Ton	Converter initialization time	ADCCx(2) changes from 0 to 1, LDKx, VDD or VB measurements			500	μs
702	Tconv	Conversion time				140	μs
703	R(ADC)	A/D Converter Resolution				10	bit
704	RAC	Relative Accuracy		-1		+1	LSB
705	VZS()	Zero Scale Voltage	ADCx(9:0) = 000h		0		V
706	VFS()	Full Scale Voltage	ADCx(9:0) = 3FFh	1.0	1.1	1.2	V
707	MDAM	MDAx Measurement	MDAx = 0.5 V, ADCCx(2:0) = 100, ADSNFx = 0	372	465	558	LSB
708	VDDM	VDD Measurement	VDD = 3.3 V, ADCCx(2:0) = 110	312	390	468	LSB
709	VBM	VB Measurement	VB = 8 V, ADCCx(2:0) = 101	744	930	1023	LSB
710	LDKM	LDKx Measurement	LDKx = 8 V, ADCCx(2:0) = 111	744	930	1023	LSB
Overt	emperature						
B01	Toff	Overtemperature Shutdown	Rising temperature	130		170	°C
B02	Ton	Overtemperature Release	Falling temperature	120		165	°C
B03	Thys	Hysteresis	Toff – Ton	3			°C
Temp	erature Mon	itor					
C01	Trange	Temperature Measurement Range		-40		125	°C
C02	Tresol	Temperature Measurement Reso- lution			1		°C
C03	Reading	Temperature Value Ranges	Tj = 125 °C Tj = -40 °C	160 0		190 15	digits digits
DCO	Output						
D01	lsc()hi	DCO Output Current	V(VDD) = 33.5 V, V(DCO) < 1.4 V, RDCO = 0x3F	-175	-130	-85	μA
D02	lleak	Leakage Current at DCO	RDCO = 0x00 or NSTBY = lo, V(DCO) = 0 5.5 V	-1		1	μA
D03	I(DCO)LSB	I(DCO) Resolution	V(DCO) < 1.4 V	1.3	2	2.7	μA
Oscill	ator						
E01	Fosc	Oscillator Frequency	NSTBY = hi	100	200	400	kHz
E02	T(cfgtimo)	Configuration Mode Timeout	MODE(1:0) = 10	40	82	164	ms
E03	tWDT	Watchdog Timeout	NSTBY = hi	20		120	μs

#### **OPERATING REQUIREMENTS: SPI and I<sup>2</sup>C Interface**

Operating Conditions: VB = 2.8 ... 11 V, Tj = -40 ... 125 °C

ltem	n Symbol Parameter Conditions		Conditions		Unit	
No.	-			Min.	Max.	
SPI / I <sup>2</sup> C Interface Timing						
1001		Setup Time: NCS/A1 hi $\rightarrow$ lo before SCLK lo $\rightarrow$ hi	INS/WKR = Io	20		ns
1002		Setup Time: MOSI/A0 stable before SCLK/SCL Io $\rightarrow$ hi	INS/WKR = Io	20		ns

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#### **OPERATING REQUIREMENTS: SPI and I<sup>2</sup>C Interface**

Opera	ting Conditi	ons: VB = 2.8 11 V, Tj = -40 125 °C				
ltem No.	Symbol	Parameter	Conditions	Min.	Max.	Unit
1003	thDCL	Hold Time: MOSI/A0 stable after SCLK/SCL Io $\rightarrow$ hi	INS/WKR = Io	20		ns
1004	tCLh	Signal Duration SCLK/SCL hi	INS/WKR = Io	50		ns
1005	tCLI	Signal Duration SCLK/SCL lo	INS/WKR = Io	50		ns
1006	thCLC	Hold Time: NCS/A1 lo after SCLK/SCL hi $\rightarrow$ lo	INS/WKR = Io	20		ns
1007	tCSh	Signal Duration NCS/A1 hi	INS/WKR = Io	50		ns
1008	tpCLD	Propagation Delay: MISO/SDA stable after SCLK/SCL hi $\rightarrow$ lo	INS/WKR = Io, V(VDD) > 3 V, Cload = 10 pF, no external pull-up	0	30	ns
1009	tHIZ	MISO to HIZ delay	INS/WKR = Io	0	25	ns
1010	f(SCLK)	SPI clock frequency			10	MHz
I011	f(SCL)	I <sup>2</sup> C clock frequency			400	kHz



Figure 1: SPI interface timing

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#### **OPERATING AND STANDBY MODES**

iC-HT operating modes are selected by the pins EMC and INS. These pins are intended to be static and should not be changed during operation to avoid misconfiguration or unintended operation. It is recommended to tie the pins EMC and INS to VDD or GND. Floating pins EMC and INS/WKR are detected as faulty configuration and signaled at NCHK.

#### iC-WK mode

In iC-WK mode the iC-HT operates as an Automatic Power Control (APC) laser controller, similar to iC-Haus' iC-WK/iC-WKL, iC-WKM and iC-WKN. iC-WK mode is set by pin configuration and the external monitor resistor. Pin EMC is set to lo and pin INS/WKR selects the reference voltage:

- INS/WKR = hi for 0.5 V reference voltage, similar to iC-WK, iC-WKL, iC-WKN.
- INS/WKR = lo for 0.25 V reference voltage, similar to iC-WM.

Floating pins EMC and INS/WKR are detected as faulty configuration and signaled at NCHK.

#### MCU mode

In microcontroller unit (MCU) mode, iC-HT features two control modes: automatic power control (APC) and automatic current control (ACC). Pin EMC is set to hi and pin INS/WKR selects the serial communication interface protocol. Selection of the communication protocol is achieved through pin INS/WKR:

- INS/WKR = hi for I2C.
- INS/WKR = Io for SPI.

Floating pins EMC and INS/WKR are detected as faulty configuration and signaled at pin NCHK. Any input interface could be enabled if EMC or INS pins are open. Comunication with the chip might be possible but laser cannot be switched on. This situation is only signaled at pin NCHK, which would remain low as long as any of the EMC or INS pins are unconnected. MCU must monitor the status of the pin NCHK to get all the status information of the chip.

#### Standby mode

iC-HT in standby mode has a very low current consumption (< 10  $\mu$ A) and does retain its configuration. Standby mode will not reset the internal RAM.

In order to exit standby mode, pin NSTBY must be set to hi (e.g. VB). VDD is switched off in standby mode and can not be used to exit standby mode.

VB, NSTBY and LDKx withstand voltages up to 11 V, whereas the remaining input pins operate up to 5.5 V and do have high impedance at standby mode.

Information on timing after waking up from standby mode can be found on page 36.

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#### LASER DIODE/LED TYPES AND OPERATION MODES

For APC operation a monitor diode is required. This operation is possible in microcontroller unit (MCU) mode and in the iC-WK mode.

iC-HT can operate in APC two types of laser diodes/LEDs with monitor diodes:

- N-type laser diodes
- M-type laser diodes



Figure 2: iC-HT with N-type laser diode

In automatic current control (ACC) operation there is no monitor diode and any diode can be operated in microcontroller unit (MCU) mode.

All operations are possible with laser diodes (LDs) or light emitting diodes (LEDs). In the following text we do not differ between laser diodes (LDs) and light emitting diodes (LEDs).



Figure 3: iC-HT with M-type laser diode

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#### **iC-WK MODE**

must be cycled.

Setting pin EMC = Io configures iC-HT for iC-WK mode. EMC pin must be set to GND.

In iC-WK mode both channels operate in APC mode. The internal programmable logarithmic monitor resistors are disabled, therefore connection of external resistors at pins MDAx is required.

For re-enabling the channel, the corresponding ECx pin

sistors at pins MDAx is required. The APC reference can be set to two different values by means of pin INS/WKR, as explained in table 1, and the overcurrent threshold is set to its maximum value of 750 mA (cf. *Electrical Characteristics No. 107*). In case of overcurrent, the respective channel is disabled.

Reference Voltage in iC-WK mode					
INS/WKR Reference similar to Voltage					
Lo	0.25 V	iC-WKM			
Hi 0.5 V		iC-WK, iC-WKL, iC-WKN			

Table 1: Reference selection (cf. *Electrical Characteristics No. 110*)

External CI capacitors must be added in this operation mode at pins CIx and CILx. Figure 4 shows an example in iC-WK mode using an N-type laser diode, where 0.5 V reference is selected. Figure 5 presents the same configuration with an M-type laser diode.



Figure 4: iC-HT in iC-WK mode with N-type laser diode

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Laser Diodes



# **iC-HT** DUAL CW LASER DIODE DRIVER



Figure 5: iC-HT in iC-WK mode with M-type laser diode

In the configuration as per Figures 4 and 5 pin, NSTBY is connected to VB. This is required to force iC-HT leaving standby mode and starting normal operation, as the pin includes an internal pull-down resistor.

#### Laser channel enabling

Setting pins EC1, EC2 to hi enables the corresponding channels. In order to ensure safe operation of iC-HT, several events automatically disable both output channels:

- Pins INS/WKR or EMC left unconnected (IN-SOPEN, EMCOPEN), iC-HT enters error mode and the laser channels cannot be enabled.
- Supply power-down either at VB (PDOVB) or VDD (PDOVDD), the laser channels are unconditionally disabled during the power down event.

 Overcurrent (OVC) or overtemperature (OVT), laser channels are switched off. Cycling pins EC1, EC2 or a power-up is required to switch on the laser again.

These events are signalled with a low level at pin NCHK (see Fig. 6)



Figure 6: iC-HT in iC-WK mode, NCHK function

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R/W 000

#### MICROCONTROLLER MODE

Setting pin EMC to hi configures iC-HT for microcontroller mode (MCU mode). Pin EMC must be set using a pull-up resistor or directly short-circuited to pin VDD. Several parameters can be configured through a microcontroller via I<sup>2</sup>C or SPI communication. See page 24.

The configuration of the internal parameters of iC-HT must be done in configuration mode. In this mode, the configuration memory can be written and read back without changing the previous configuration state of iC-HT. Once the configuration is considered as valid, iC-HT can be switched to operation mode. These two modes are configured by the MODE register. If the time in configuration mode exceeds the Configuration Mode Timeout (cf. *Electrical Characteristics No. E02*), both channels will be switched off. See page 34.

Each individual channel can be enabled by setting pin ECx to hi. Setting register bits DISCx to 1 disables the corresponding channel. If either pin ECx is lo or register bits DISCx is 1, the corresponding channel is disabled.

Addr. 0x10; bit 3

Code	Function
0xx	Disabled
100	V(MDA1), ADSNF1 = 0
100	V(PLR1), ADSNF1 = 1
101	V(VB)
110	V(VDD)
111	V(LDK1)

Addr. 0x10; bit 7:5

ADCC1(2:0)

Table 4: ADC channel 1 source selection

ADCC2(2:0)	Addr. 0x15; bit 7:5	R/W 000
Code	Function	
0xx	Disabled	
100	V(MDA2), ADSNF2 = 0	
100	V(PLR2), ADSNF2 = 1	
101	V(VB)	
110	V(VDD)	
111	V(LDK2)	

Table 5: ADC channel 2 source selection

 Function

 Channel 1 can be enabled by pin EC1

 Channel 1 cannot be enabled by pin EC1

 Table 2: Disable channel 1

R/W 1

DISC2	Addr. 0x15; bit 3	R/W 1
Code	Function	
0	Channel 2 can be enabled by pin EC2	
1	Channel 2 cannot be enabled by pin EC2	

Table 3: Disable channel 2

Different voltages can be measured using a 10 bit A/D converter with two resolutions. The following internal voltages can be measured:

- V(LDKx) up to 8 V with 8.6 mV resolution
- V(VDD) up to 8 V with 8.6 mV resolution
- V(VB) up to 8 V with 8.6 mV resolution
- V(MDAx) up to 1 V with 1.075 mV resolution
- · V(PLRx) up to 1 V with 1.075 mV resolution

The register bits ADCCx select the signal measured with the 10 bit A/D converter.

With ADCCx(2:0) = 100, the signal to the A/D converter is selected by register bit ADSNFx. With ADSNFx = 1 the measuring point to the A/D converter is the internal sense node of the internal programmable logarithmic monitor resistor (PLR). With ADSNFx = 0 the sensing point is connected directly to MDAx pin.

ADSNF1	Addr. 0x1A; bit 2	R/W 0			
Code	Function				
0	ADC measurement pin MDA1 (force)	ADC measurement pin MDA1 (force)			
1	ADC measurement PLR1 (sense)				

Table 6: ADC channel 1 sense/force selection

ADSNF2	Addr. 0x1A; bit 6	R/W 0
Code	Function	
0	ADC measurement pin MDA2 (force)	
1	ADC measurement PLR2 (sense)	
	·	

Table 7: ADC channel 2 sense/force selection

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DISC1

Code

0





Figure 7: Channel 1 schematic



Figure 8: APC mode simplified

Two different control modes can be configured independent for each channel: automatic power control (APC) and automatic current control (ACC). In both modes a 10 bit logarithmic D/A converter sets the reference voltage and an 8 bit programmable D/A converter configures the overcurrent threshold.

#### APC (Automatic Power Control) mode

In APC mode the laser power is controlled. The monitor diode current is used as feedback in the laser power control loop. APC mode is selected by setting EACCx register bit to 0.

EACC1	Addr. 0x10; bit 0	R/W 0
Code	Function	
0	APC mode enabled for channel 1	
1	ACC mode enabled for channel 1	

Table 8: APC/ACC in channel 1

EACC2	Addr. 0x15; bit 0	R/W 0
Code	Function	
0	APC mode enabled for channel 2	
1	ACC mode enabled for channel 2	

Table 9: APC/ACC in channel 2

An example of APC with default configuration is shown in Figure 8.

An internal 8 bit programmable logarithmic monitor resistor (PLR) can be used in APC mode. In APC mode it is also possible to use an external monitor resistor connected to pin MDAx. If register bit DISPx is 0, the PLR is present. If DISPx is 1, the PLR is disabled and an external monitor resistor must be used.

DISP1	Addr. 0x10; bit 2	R/W 0
Code	Function	
0	PLR enabled for channel 1	
1	PLR disabled for channel 1	

Table 10: Disable PLR channel 1

DISP2	Addr. 0x15; bit 2	R/W 0
Code	Function	
0	PLR enabled for channel 2	
1	PLR disabled for channel 2	

Table 11: Disable PLR channel 2

Both programmable logarithmic monitor resistors (PLR) feature a wide logarithmic resistor range from  $100 \Omega$  to  $500 \text{ k}\Omega$ , in steps of typically 3.3%. This covers a wide range of monitor currents. More information about the PLR can be found on page 26.

For fine-tuning the optical power, the reference voltage can be set with a 10 bit logarithmic D/A converter, which is configurable through the register REFx. This converter has a voltage range that goes typically from 0.1 V to 1.1 V, allowing an operation resolution of typically 0.235%. See page 27.

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For calculating the minimum value of Imon, Vref(0x00, max. value) (cf. *Electrical Characteristics No.* 303) and Rmda(RMDx = 0xFF, min. value) (cf. *Electrical Characteristics No.* 201) are used.

 $Imon(min) = \frac{Vref(0x000,max)}{Rmda(RMDx=0xFF,min)} = \frac{0.11}{350000} = 0.31 \, uA$ 

It is not recommended to configure iC-HT to have such small Imon values, otherwise the leakage current at MDAx may have an influence (cf. *Electrical Characteristics No. 204*), especially at high temperatures. To avoid this, Imon should be much greater than the leakage current.

For calculating the maximum value of Imon, Vref(0x3FF, min. value) (cf. *Electrical Characteristics No. 303*) and Rmda(RMDx = 0x00, max. vaule) (cf. *Electrical Characteristics No. 201*) are used. The following formula can to be used for calculating Rmda(RMDx = 0x00, max. value):

 $Rmd = Rmd_0(1 + \frac{\Delta Rmd(\%)}{100})^{n+1}$ , n from 0 to 255

 $Rmda(RMDx = 0x00, max) = Rmd_0(1 + \frac{\Delta Rmd(\%)}{100})^{16}$ 

 $286 = Rmd_0(1 + \frac{3.3}{100})^{16}$ 

 $Rmd_0 = 170\Omega$ 

Therefore:

 $Imon(max) = \frac{Vref(0x3FF,min)}{Rmd_0} = \frac{1.00}{170} = 5.88 \, mA$ 

Any other Imon value can be calculated using Rmd formula above. Due to its logarithmic characteristic, the steps between two consecutive values is kept within 3.3 % typical value.

The programmable overcurrent shutdown can be set to protect the laser by disabling the channel. The overcurrent threshold is configurable in two different ranges. The range is selected through register bit RACCx. If RACCx = 1, the overcurrent threshold is in the low range, up to 90 mA. If RACCx = 0, the over-current threshold is in the high range, up to 750 mA (cf. *Electrical Characteristics No. 107*).

RACC1	Addr. 0x1A; bit 0	R/W 0
Code	Function	
0	Current range high for channel 1	
1	Current range low for channel 1	

Table 12: RACC1 current range configuration channel

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RACC2	Addr. 0x1A; bit 4	R/W 0
Code	Function	
0	Current range high for channel 2	
1	Current range low for channel 2	

Table 13: RACC2 current range configuration channel 2

In each range, the particular overcurrent threshold value can be set in register ILIMx. The regulator does not differ the two current ranges in APC.

Note: The accuracy of the overcurrent detection depends on the correct biasing of the driver stage. If the driver is saturated, the accuracy of the overcurrent detection is reduced and may not work as expected. Saturation voltage (*Electrical Characteristics No. 101*) must be ensured for a proper function of the overcurrent detection. LDKSAT1 and LDKSAT2 status bits can be used for anticipated saturation detection.

ILIM1	Addr. 0x11; bit 7:0	R/W 0xFF
Code	Function	
0x0A	Channel 1 overcurrent threshold set current	to minimum
	Channel 1 overcurrent threshold set $llim = (\Delta l(LDK) \cdot n)$ , n from 10 to 25.	
0xFF	Channel 1 overcurrent threshold set current	to maximum



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ILIM2	Addr. 0x16; bit 7:0	R/W 0xFF
Code	Function	
0x0A	Channel 2 overcurrent threshold current	set to minimum
	Channel 2 overcurrent threshold $llim = (\Delta I(LDK) \cdot n)$ , n from 10 to	
0xFF	Channel 2 overcurrent threshold current	set to maximum

Table 15: Overcurrent threshold configuration channel 2

An overcurrent event can be simulated using SOVCx. If SOVCx = 1, the corresponding overcurrent error bit OVCx will be set to 1, the error will be signaled at NCHK and the corresponding laser channel will be disabled. The overcurrent error will remain forced until SOVCx = 0.

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SOVC1	Addr. 0x1D; bit 5	R/W 0
Code	Function	
0	No Overcurrent event at channel 1 is	simulated.
1	Overcurrent event at channel 1 simula	ited.

Table 16: Simulate overcurrent channel 1

SOVC2	Addr. 0x1D; bit	6 R/W 0
Code	Function	
0	No overcurrent event at channel 2 is simulated.	
1	Overcurrent event at channel 2 simulated.	

Table 17: Simulate overcurrent channel 2

#### ACC (Automatic Current Control) mode

In this mode, the laser diode current is controlled and no monitor diode is required. ACC mode is selected setting EACCx register bit to 1.

EACC1	Addr. 0x10; bit 0	R/W 0
Code	Function	
0	APC mode enabled for channel 1	
1	ACC mode enabled for channel 1	

Table 18: APC/ACC in channel 1

EACC2	Addr. 0x15; bit 0	R/W 0
Code	Function	
0	APC mode enabled for channel 2	
1	ACC mode enabled for channel 2	

Table 19: APC/ACC in channel 2

In ACC mode the internal programmable logarithmic monitor resistor (PLR) is not used. Instead the internal RACCx resistor is used to set the current. For fine-tuning the regulated current, the reference voltage can be set with a 10 bit logarithmic D/A converter, which is configurable through the register REFx. This converter has a voltage range that goes typically from 0.1 V to 1.1 V, allowing an operation resolution of typically 0.235%. See page 27. Figure 9 shows an example of this configuration.

Two different current ranges can be set through register bit RACCx, with RACCx = 1 for up to 75 mA and with RACCx = 0 for up to 650 mA. The programmable overcurrent shutdown protects against over regulations during the laser power-on. Table 20 shows some typical current settings. For detailed limits, please refer to *Electrical Characteristics No. 114* 

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ACC typical current settings			
REFx	RACCx=0	RACCx=1	
0x000	77.0 mA	9.06 mA	
0x001	77.2 mA	9.08 mA	
0x010	77.4 mA	9.10 mA	
0x200	256.2 mA	30.14 mA	
0x201	256.8 mA	30.21 mA	
0x202	257.4 mA	30.28 mA	
0x3FD	846.0 mA	99.53 mA	
0x3FE	848.0 mA	99.77 mA	
0x3FF	850.0 mA	100 mA	

Table 20: ACC typical current settings



Figure 9: ACC mode simplified

RACC1	Addr. 0x1A; bit 0	R/W 0
Code	Function	
0	Current range high for channel 1	
1	Current range low for channel 1	

Table 21: RACC1 configuration

RACC2	Addr. 0x1A; bit 4	R/W 0
Code	Function	
0	Current range high for channel 2	
1	Current range low for channel 2	

Table 22: RACC2 configuration

In each range, the particular overcurrent threshold value can be set in register ILIMx. The regulator does differ the two current ranges in ACC.

Note: The accuracy of the overcurrent detection depends on the correct biasing of the driver stage. If the driver is saturated, the accuracy of the overcurrent

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detection is reduced and may not work as expected. Saturation voltage (*Electrical Characteristics No. 101*) must be ensured for a proper function of the overcurrent detection. LDKSAT1 and LDKSAT2 status bits can be used for anticipated saturation detection.

ILIM1	Addr. 0x11; bit 7:0	R/W 0xFF
Code	Function	
0x0A	Channel 1 overcurrent threshold se current	et to minimum
	Channel 1 overcurrent threshold set $llim = (\Delta l(LDK) \cdot n)$ , n from 10 to 2	
0xFF	Channel 1 overcurrent threshold se current	et to maximum

Table 23: Overcurrent threshold configuration channel

ILIM2	Addr. 0x16;	bit 7:0	R/W 0xFF
Code	Function		
0x0A	Channel 2 overcurre current	ent threshold set to	minimum
	Channel 2 overcurre $llim = (\Delta l(LDK) \cdot n)$		
0xFF	Channel 2 overcurre current	ent threshold set to	maximum

Table 24: Overcurrent threshold configuration channel 2

An overcurrent event can be simulated using bit SOVCx. If SOVCx = 1, the corresponding overcurrent error bit OVCx will be set to 1, the error will be signaled through NCHK and the corresponding laser channel will be disabled. The overcurrent error will remain forced until SOVCx = 0.

SOVC1	Addr. 0x1D; bit 5	R/W 0
Code	Function	
0	No overcurrent event at channel 1 is sim	
1	Overcurrent event at channel 1 simulate	d.

Table 25: Simulate overcurrent channel 1

SOVC2	Addr. 0x1D; bit 6 R/	/W 0
Code	Function	
0	No overcurrent event at channel 2 is simulated	i.
1	Overcurrent event at channel 2 simulated.	

Table 26: Simulate overcurrent channel 2

An external capacitor can be added in ACC mode in order to avoid oscillations as it is shown in Figure 9. The external CIx must be enabled setting the bit ECIEx (Tables 29 and 30). Rev C1, Page 18/46

In ACC mode, the MDAx pin can be monitored through a 10 bit A/D converter. This can be used for measuring the laser light power, if a photodiode is connected to pin MDAx, as it is shown in Figure 10. This allows adjusting the voltage reference in order to set the laser current and obtain the desired laser light power.

The internal programmable logarithmic monitor Resistor (PLR), if enabled (DISPx = 0), gives feedback for the current control through the 10 bit A/D converter. Register bit ADSNFx must be set to 1 in order to measure the internal sense node. An external monitor resistor can be used to measure the optical power, achieved by setting DISPx to 1. Therefore register bit ADSNFx must be set to 0 in order to measure directly at pin MDAx.



Figure 10: ACC with monitor photodiode

ACC mode permits to combine both channels in one iC-HT (see chapter COMBINING BOTH CHANNELS) and several iC-HT in parallel. When both channels are combined the programmable overcurrent shutdown is by channel. Another option is to connect the LDK together when both channels are configured in ACC mode and the channel 1 with high current range configuration and the channel 2 with low current range configuration. With this pre-sets there is a granularity about 0.2% using the channel 1 steps regulation and about 0.03% using the channel 2 in the regulation.

#### The Regulator

In MCU mode the control can be carried out without the need of external capacitor. This allows a fast response of the regulator. The speed of the regulator's response and stability can be configured using three bits (COMPx), providing a compensation factor.

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COMP1	Addr. 0x13; bit 6:4 R/W 011
Code	Function
000	Minimum compensation for the channel 1 regulator slower response of regulator
	Maximum compensation for the channel 1 regulato
111	faster response of regulator

Table 27: Regulator delay compensation channel 1

COMP2	Addr. 0x18;	bit 6:4	R/W 011
Code	Function		
000	Minimum compensa slower response of r		channel 2 regulator,
111	Maximum compensa faster response of re		e channel 2 regulator,

Table 28: Regulator delay compensation channel 2

Alternatively it is possible to use external capacitors connected to pins Clx and ClLx. In this case, register bit ECIEx should be set to 1 and COMPx to its highest value, "111".

ECIE1	Addr. 0x10; bit 1	R/W 0
Code	Function	
0	External CI capacitor for channel 1 disc	connected
1	External CI capacitor for channel 1 con	inected

Table 29: Enable external capacitor channel 1

ECIE2	Addr. 0x15; bit 1	R/W 0
Code	Function	
0	External CI capacitor for channel 2 d	
1	External CI capacitor for channel 2 c	onnected

Table 30: Enable external capacitor channel 2

The regulator is offset compensated in order to prevent optical power drifts. Offset compensation can be disabled by setting register bit EOCx to 0.

EOC1	Addr. 0x10; bit 4	R/W 1
Code	Function	
0	Channel 1 regulator offset compensation	n disabled
1	Channel 1 regulator offset compensation	n enabled

Table 31: Enable offset compensation channel 1

EOC2	Addr. 0x15; bit 4	R/W 1
Code	Function	
0	Channel 2 regulator offset compen	sation disabled
1	Channel 2 regulator offset compen	sation enabled

Table 32: Enable offset compensation channel 2

An internal oscillator is used for the offset compensation. A watchdog timer (WDT) is included in order to monitor proper function of the oscillator. If an error is detected by the WDT, the laser channels are disabled, OSCERR error bit is set in STATUS0 register and the error event is signaled at pin NCHK. This error signaling can be suppressed using the mask register bit MOSCERR (set to 1).

MOSCERR	Addr. 0x1D; bit 0	R/W 0
Code	Function	
0	Oscillator error (watchdog) will be sign	
1	Oscillator error (watchdog) will not be NCHK	signaled at

Table 33: Oscillator error mask

iC-HT monitors the saturation voltage of the regulator's output transistor at pin LDKx. The LDK saturation threshold can be configured through register bits RLD-KSx.

		D/14/00
RLDKS1	Addr. 0x13; bit 3:2	R/W 00
Code	Function	
00	V(LDK1) < 0.5 V sets the LDKSAT1 alar	m bit
01	V(LDK1) < 0.8 V sets the LDKSAT1 alar	m bit
10	V(LDK1) < 1.0 V sets the LDKSAT1 alar	m bit
11	V(LDK1) < 1.2 V sets the LDKSAT1 alar	m bit

Table 34: LDK saturation threshold selection channel 1

RLDKS2	Addr. 0x18; bit 3:2	R/W 00
Code	Function	
00	V(LDK2) < 0.5 V sets the LDKSAT2 alarm	
01	V(LDK2) < 0.8 V sets the LDKSAT2 alarm	bit
10	V(LDK2) < 1.0 V sets the LDKSAT2 alarm	bit
11	V(LDK2) < 1.2 V sets the LDKSAT2 alarm	bit

Table 35: LDK saturation threshold selection channel 2

If the LDKx voltage falls below the LDK saturation threshold the LDKSATx error bit in STATUS1 register will be set and it will be signaled through output pin NCHK. Setting the mask register bit MLDKSATx to 1 suppresses the signaling at NCHK.

MLDKSAT1	Addr. 0x1D; bit 2	R/W 1
Code	Function	
0	LDKSAT1 event will be signaled at NC	
1	LDKSAT1 event will not be signaled a	t NCHK

Table 36: LDK saturation mask channel 1

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MLDKSAT2	Addr. 0x1D; bit 3	R/W 1
Code	Function	
0	LDKSAT2 event will be signaled at NCHK	
1	LDKSAT2 event will not be signaled at NCI	нκ

Table 37: LDK saturation mask channel 2

#### Laser channel enabling and error handling

With pin INS/WKR or EMC unconnected, a corresponding error signal will be generated (INSOPEN, EM-COPEN) and will disable the laser channels. Any input interface could be enabled if EMC or INS pins are open. Comunication with the chip might be possible but laser

cannot be switched on. This situation is only signaled at the MCU interruption pin NCHK, which would remain low as long as any of the EMC or INS pins are unconnected. MCU must monitor the status of the pin NCHK to get all the status information of the chip

Setting DISC1 and DISC2 to 1(default) disables the corresponding channel.

The errors in STATUS0 and STATUS1 registers disable the laser channels. Every change in the STATUS registers is signaled at pin NCHK, unless the error event is masked by the corresponding error mask bit.

Register	Address	Bits	Default	Description
INITRAM	0x00	0	R/O	RAM initialized.
PDOVDD	0x00	1	R/O	Power down event at VDD
MEMERR	0x00	2	R/O	RAM memory validation error
OVT	0x00	3	R/O	Overtemperature event
OVC2	0x00	4	R/O	Overcurrent at channel 2
OVC1	0x00	5	R/O	Overcurrent at channel 1
OSCERR	0x00	6	R/O	Oscillator error (watchdog set)
CFGTIMO	0x00	7	R/O	Configuration mode timeout event
MAPC1	0x01	0	R/O	Channel 1 current state (on or off)
MONC1	0x01	1	R/O	Monitor channel 1 enabled at least once (latched)
LDKSAT1	0x01	2	R/O	Channel 1 LDK saturation event
MAPC2	0x01	4	R/O	Channel 2 current state (on or off)
MONC2	0x01	5	R/O	Monitor channel 2 enabled at least once (latched)
LDKSAT2	0x01	6	R/0	Channel 2 LDK saturation event

Table 38: Status registers overview

In order to enable the channels, the error events must be acknowledged. Acknowledging an error is accomplished by reading the STATUS register. After a power-on PDVDD and INITRAM errors will be set, therefore it is required to read STATUS0 and STATUS1 registers after each power-on.

Exiting standby mode will not reset the RAM but will set the PDOVDD status bit. Therefore STATUS0 must be read once after each standby to re-enable the laser channels.

In case of an overcurrent (OVC) or an overtemperature (OVT) event, laser channels are disabled.

A memory error event and a configuration timeout error event will also disable the laser channels. See page 34. The conditions to enable each laser channel are shown in Figure 11.

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# **iC-HT** DUAL CW LASER DIODE DRIVER



Figure 11: Laser control logic in MCU mode

The events that require user attention are signalled as a low level at pin NCHK. Some of these events are considered as non-critical and can be masked. Figure 12 shows the logical function of NCHK. As a safety feature, some critical events (Laser overcurrent, system overtemperature, local oscillator malfunction) can be simulated by the user. This eases the debugging of applications where the reaction in case of critical events is important.



Figure 12: NCHK function in MCU mode

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#### COMBINING BOTH CHANNELS

iC-HT can drive one laser diode up to 1500 mA with both channels combined.

Therefore register bit MERGE must be set to 1. Disable channel register bits DISC1 and DISC2 must both be set to 0 and both enable channel pins EC1 and EC2 must be set hi.

MERGE	Addr. 0x1B; bit 6	R/W 0
Code	Function	
0	Channel 1 and 2 operate independently	
1	Power transistor from channel 2 in para channel 1, controlled by channel 1	llel with

Table 39: Channel merging

When both channels are combined the control is done by channel 1. APC and ACC can both be used with both channels combined. In ACC mode, the reference needs be set to 50% of the desired current value. This is not required for APC.

REF1 RACCx=0 RACCx=			
	RMERGE=1	MERGE=1	
0x000	144.0 mA	18.12 mA	
0x001	144.4 mA	18.16 mA	
0x010	144.8 mA	18.20 mA	
0x200	512.4 mA	60.28 mA	
0x201	513.6 mA	60.42 mA	
0x202	514.8 mA	60.56 mA	
0x3FD	1692.0 mA	199.54 mA	
0x3FE	1696.0 mA	199.54 mA	
0x3FF	1700.0 mA	200 mA	

Table 40: ACC typical current settings with MERGE = 1



Figure 13: Combining both channels

In combined mode, the internal CI nodes of both channels are connected together. If required, an external capacitor on channel 1 can be used for improved stability. It is possible to have an additional external capacitor on channel 2, if ECIE2 bit is set to 1. Having an exter-

nal capacitor on channel 2 and none on channel 1 is invalid.

On combined operation both overcurrent thresholds are active. The overcurrent threshold channel 1 needs to

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be set to to 50% and the overcurrent threshold channel 2 should be disabled by setting it to its maximum value (0xFF). An overcurrent will only be detected on channel 1.

ILIM1	Addr. 0x11; bit 7:0 R/W 0xFF	
Code	Function	
0x0A	Channel 1 overcurrent threshold: 80 mA, MERGE=	1
	Channel 1 overcurrent threshold set to $llim = (2 * \Delta l(LDK) \cdot n)$ , n from 10 to 255, MERGE=	1
0xFF	Channel 1 overcurrent threshold: 2040 mA, MERGE=1	

Table 41: Overcurrent threshold configuration channel 1 with MERGE = 1





Figure 14: Additional photodiode in combined configuration

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#### SERIAL COMMUNICATION INTERFACES

#### SPI slave interface

The SPI slave interface is enabled by setting pin INS/WKR to Io and uses pins NCS/A1, SCLK/SCL, MISO/SDA and MOSI/A0. Pin NCS/A1 is the chip select pin and must be set Io by the SPI master in order to start communication. Pins MISO/SDA and MOSI/A0 are the data communication lines and pin SCLK/SCL is the clock line generated by the SPI master (e.g. microcontroller). The SPI protocol frames are shown in Figure 15.

A communication frame consists of one address byte and at least one data byte. Bits 7:6 of the address byte is the opcode used for selecting a read operation (set to "10") or a write (set to "01") operation. The remaining 6 bits are used for register addressing.

It is possible to transmit several bytes consecutively, if the NCS signal is not reset and SCLK/SCL keeps clocking, as it is shown in Figure 15. The address is internally incremented after each transmitted byte. Once the address reaches the last register (0x3F), it is reset back to 0x00.



#### I<sup>2</sup>C slave interface

(see Table 42 and 43).

The I<sup>2</sup>C slave interface is enabled by setting pin INS/WKR to hi and uses pins NCS/A1, SCLK/SCL, MISO/SDA and MOSI/A0. The protocol frames are shown in Figure 16.

Action	b7	b6	b5	b4	b3	b2	b1	b0
Write to slave	1	1	0	0	0	A1	A0	0
Read from slave	1	1	0	0	0	A1	A0	1

A communication frame consists of one slave address byte, one register address byte and at least one data byte. Bits 7:1 of the slave address byte form the slave

byte, one register address byte and at least one data byte. Bits 7:1 of the slave address byte form the slave identification code (ID) and bit 0 is used for specification of the data direction (0 for write, 1 for read). The slave ID consists of 7 bits. The five most significant bits are fixed by default to value 0b11000. Pins MOSI/A0 and NCS/A1 are used to set the remaining slave ID bits

#### Table 42: I<sup>2</sup>C write/read byte

Action	A1	A0	Slave ID	Address byte
Write to slave 0	lo	lo	0x60	0xC0
Read from slave 0	lo	lo	0x60	0xC1
Write to slave 1	lo	hi	0x61	0xC2
Read from slave 1	lo	hi	0x61	0xC3
Write to slave 2	hi	lo	0x62	0xC4
Read from slave 2	hi	lo	0x62	0xC5
Write to slave 3	hi	hi	0x63	0xC6
Read from slave 3	hi	hi	0x63	0xC7

Table 43: I<sup>2</sup>C write/read address

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I2C READ COMMAND FRAME

Figure 16: I<sup>2</sup>C read and write commands

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#### 8 BIT INTERNAL PROGRAMMABLE LOGARITHMIC MONITOR RESISTORS

In MCU mode internal 8 bit programmable logarithmic monitor resistors (PLRx) are provided for APC.

The resistor value can be selected from 256 values, ranging from  $100 \Omega$  to  $500 \text{ k}\Omega$ , following logarithmic increments with a typical step width of 3.3%. The resistors are configured with registers RMDx(7:0).

RMD1	Addr. 0x12; bit 7:0	R/W 0xFF
Code	Function	
0x00	PLR1 set to the minimum resistant	се
	PLR1 set to $Rmd = Rmd_0(1 + \frac{\Delta Rmd(\%)}{100})^{n+1}$ , n	from 0 to 255
0xFF	PLR1 resistor set to the maximum	resistance

Table 44: MDA resistor channel 1

RMD2	Addr. 0x17; bit 7:0	R/W 0xFF
Code	Function	
0x00	PLR2 resistor set to the minimum re	esistance
	PLR2 set to $Rmd = Rmd_0(1 + \frac{\Delta Rmd(\%)}{100})^{n+1}$ , n fr	om 0 to 255
0xFF	PLR2 resistor set to the maximum r	esistance

Table 45: MDA resistor channel 2

The following formula calculates the register RMDx in order to set the desired resistor value:

 $Rmd = Rmd_0(1 + \frac{\Delta Rmd(\%)}{100})^{n+1}$ , n from 0 to 255

Where  $Rmd_0$  is the minimum resistor value (typically 100  $\Omega$ ),  $\Delta Rmd(\%)$  is the step between two consecutive resistor values (typically 3.3%) and *n* is the value of RMDx register in decimal.

In APC mode the regulation node is the internal connection to PLR, it is not MDAx pin (see detail in Figure 17). Voltage present at pin MDAx may differ from the internal regulation node. This regulation node can be sensed through the 10 bit A/D converter and read at register ADCx. Register bit ADSNFx must be set to 1 for this purpose. If ADSNFx is set to 0, MDAx pin will be the input of the A/D converter.

At pin MDAx only the 4 MSB of the RMDx configuration from PLRx are measurable. The 8 bits of the PLRx configuration RMDx can be measured with the A/D converter setting ADSNFx to 1.



Figure 17: PLR internal node regulation

With DISPx = 0 the PLRx is enabled and DISPx = 1 disables the PLRx.

DISP1	Addr. 0x10; bit 2	R/W 0
Code	Function	
0	PLR enabled for channel 1	
1	PLR disabled for channel 1	

Table 46: Disable PLR channel 1

DISP2	Addr. 0x15; bit 2	R/W 0	
Code	Function		
0	PLR enabled for channel 2		
1	PLR disabled for channel 2		

Table 47: Disable PLR channel 2

In ACC mode the PLR is not used in the control circuit. Instead, the internal RACCx resistor is used in the control loop.

Even though the PLR is not in the control circuit, it can be enabled (DISPx = 0) in order to give feedback

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through the 10 bit A/D converter for controlling the optical power, if a monitor diode is connected.



Figure 18: PLR in APC

Register bit ADSNF is set to 1 to measure the internal sense node. Alternatively, an external monitor resistor

#### **10 BIT LOGARITHMIC D/A CONVERTER**

The 10 bit logarithmic D/A converter is used for setting the regulator's voltage reference. The D/A converter is active in all operating modes. In iC-WK mode only two values are available: 0.25 V (setting INS/WKR pin lo) and 0.5 V (setting INS/WKR pin hi). In MCU mode both APC and ACC use the D/A converter. With a range from 0.1 to 1.1 V and the typical step width is 0.235%.

The D/A converter is configured through register REFx(9:0). With REFx(9:0) = 0x000, D/A output value is set to 0.1 V, and for REFx(9:0) = 0x3FF, D/A output is configured to 1.1 V.

REF1	Addr. 0x13/14; bit 9:0 R/W 0x000
Code	Function
0x000	Channel 1 regulator reference voltage set to minimum voltage
	Channel 1 regulator reference voltage set to $Vref = Vref_0(1 + \frac{\Delta Vref(\%)}{100})^{n+1}$ , n from 0 to 1023
0x3FF	Channel 1 regulator reference voltage set to maximum voltage

Table 48: Channel 1 regulator voltage reference

can be used to measure the optical power, by setting

DISPx to 1. Then register bit ADSNF must be set to 0

in order to measure directly at pin MDAx.



Figure 19: ACC with monitor photodiode

REF2	Addr. 0x18/19; bit 9:0	R/W 0x000	
Code	Function		
0x000	Channel 2 regulator reference voltage set to minimum voltage		
	Channel 2 regulator reference voltage $Vref = Vref_0(1 + \frac{\Delta Vref(\%)}{100})^{n+1}$ , n from		
0x3FF	Channel 2 regulator reference voltage	ge set to	

Table 49: Channel 2 regulator voltage reference

To calculate the D/A converter value for each REFx value, use the following formula:

*Vref* = *Vref*<sub>0</sub>(1 +  $\frac{\Delta Vref(\%)}{100}$ )<sup>*n*+1</sup>, n from 0 to 1023

Where  $Vref_0$  is the minimum value (typically 0.1 V),  $\Delta$  Vref(%) is the step value (typically 0.235%) and n is the value of REFx register in decimal.

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#### **10 BIT LINEAR A/D CONVERTER**

A 10 bit linear A/D converter is available for each channel when working in MCU mode. A variety of voltages can be measured by the converter with two resolutions:

- V(LDKx) up to 8 V with 8.6 mV resolution
- V(VDD) up to 8 V with 8.6 mV resolution
- V(VB) up to 8 V with 8.6 mV resolution
- V(MDAx) up to 1 V with 1.075 mV resolution
- V(PLRx) up to 1 V with 1.075 mV resolution

Note that when the value to be converted is higher than 8 V the A/D converter is saturated at its highest conversion value.

The register bits ADCCx select the signal measured with the 10 bit A/D converter.

ADCC1(2:0)	Addr. 0x10; bit 7:5	R/W 000	
Code	Function		
0xx	Channel 1 ADC disabled		
100	Channel 1 ADC sourced by V(MDA1), ADSNF1 = 0		
100	Channel 1 ADC sourced by V(PLR1), ADSNF1 = 1		
101	Channel 1 ADC sourced by V(VB)		
110	Channel 1 ADC sourced by V(VDD)		
111	Channel 1 ADC sourced by V(LDK1)		

Table 50: ADC channel 1 source selection

ADCC2(2:0)	Addr. 0x15;	bit 7:5	R/W 000
Code	Function		
0xx	Channel 2 ADC disabled		
100	Channel 2 ADC sourced by V(MDA2), ADSNF2 = 0		
100	Channel 2 ADC sourced by V(PLR2), ADSNF2 = 1		
101	Channel 2 ADC sou	rced by V(VE	3)
110	Channel 2 ADC sou	rced by V(VI	DD)
111	Channel 2 ADC sou	rced by V(LE	0K2)

Table 51: ADC channel 2 source selection

With ADCCx(2:0) = 100, the signal to the A/D converter is selected by register bit ADSNFx (A/D converter sense not force). With ADSNFx = 1 the measuring point to the A/D converter is the internal sense node of the internal programmable logarithmic monitor resistor (PLR). With ADSNFx = 0 the sensing point is connected directly to MDAx pin.

ADSNF1	Addr. 0x1A; bit 2	R/W 0
Code	Function	
0	ADC measurement MDA1 pin (force)	
1	ADC measurement PLR1 (sense)	

Table 52: ADC channel 1 sense/force selection

ADSNF2	Addr. 0x1A; bit 6	R/W 0	
Code	Function		
0	ADC measurement MDA2 pin (force)		
1	ADC measurement PLR2 (sense)		

Table 53: ADC channel 2 sense/force selection

When enabled, the A/D converter is continuously acquiring the signal selected by ADCCx register. The conversion time, is  $140 \,\mu s$ . Changing the source requires  $500 \,\mu s$  settling time.

In order to do a measurement, register ADCx must be read. The converter does not provide an end of conversion (EOC) bit. Instead, ADCx register contains always the value of the last conversion.

The A/D converter must be stopped before the result can be read to prevent measurement change during MSB and LSB readout. This can be done by setting the ADCCx MSB to 0. The procedure is as follows:

- 1. Set ADCCx to the desired input source with the ADC disabled (ADCCx, bit 2 set to 0).
- 2. Start ADC conversion (ADCCx, bit 2 set to 1).
- 3. Wait for the settling time (Item. 701, up to 500us).
- 4. Stop the ADC (ADCCx, bot 2 set to 0).
- 5. Read the ADC data.
- 6. If ADC input source doesn't need to be changed, repeat from 2..

ADC1	Addr. 0x03/04; bit 9:0	R
Code	Function	
0x000	ADC minimum value	
0x3FF	ADC maximum value	

Table 54: ADC channel 1

ADC2	Addr. 0x05/06; bit 9:0	R
Code	Function	
0x000	ADC minimum value	
0x3FF	ADC maximum value	

#### Table 55: ADC channel 2

The voltage corresponding to the measured digital value can be directly obtained through the following formula:

```
V(LDKx, VB, VDD) = 8 * \frac{VFS}{1023} * ADCx
```

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 $V(MDAx, PLRx) = \frac{VFS}{1023} * ADCx$ 

VFS is the fullscale voltage of the A/D converter (cf. *Electrical Characteristics No.* 706) typical 1.1 V. For a more precise measurement, the A/D converter can be calibrated by measuring a known VB voltage and calculate the VFS.

If ADSNFx = 0 the sensing point is connected directly to pin MDAx. Depending on the regulation voltage, it is possible that V(MDAx) is higher than 1.1 V. When pin MDAx is the source of the A/D converter, saturation of the converter will occur. When monitoring pin MDAx with the A/D converter, V(MDAx) must be lower than 1.1 V.

#### DC/DC CONVERTER OPTIMIZATION

iC-HT provides a 6 bit configurable current at pin DCO that can be used to trim the output voltage of a DC/DC converter.

Possible application benefits with using DCO:

- DC/DC step down operation: regulation at voltages lower than power supply
- DC/DC step up operation: regulation at voltages higher than power supply
- · Efficiency enhancement

RDCO	Addr. 0x1B; bit 5:0	R/W 0x02
Code	Function	
0x00	No current	
0x3F	130 µA typ (cf. Electrical Character	ristics No. D01)
0,01		

Table 56: Digital current output register

The proposed applications can be demonstrated with a standard DC/DC converter e.g. TPS63060DSC from Texas Instruments. This converter allows an input voltage ranging from 2.5 to 12 V and offers an output voltages from 2.5 to 8 V. It is capable of delivering up to 2 A current, depending on the output voltage. Figure 20 shows a possible configuration.



Figure 20: TPS63063 DC/DC converter from TI

#### DC/DC step down operation:

**regulation at voltages lower than power supply** The resistors R1 and R2 in the feedback path allow setting the desired output value Vout. The DC/DC converter drives Vout pin in order to yield 0.5 V at feedback pin FB. The DCO output signal from iC-HT is connected to FB pin. The Vout is controlled with the internal register RDCO from iC-HT.

The DCO current into FB node controls the voltages of the divider R1 and R2 and Vout changes in order to maintain 0.5 V at FB pin. Selecting R1 and R2 needs to consider:

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# Resistors values: R1 = R2(Vout / Vfb - 1)

- Current of the voltage divider should be high enough, in comparism to the current from the pin DCO, to offer acceptable resolution. The programmable current resolution from register RDCO is 2  $\mu$ A.
- DCO current into the voltage divider will lower Vout voltage, Vout is 8 V when no current is present at DCO.

Choosing R1 to  $100\,\text{k}\Omega,$  the value of R2 can be calculated:

$$R2 = \frac{R1}{\frac{Vout}{MR} - 1} = \frac{100k}{\frac{8}{0E} - 1} = 6.7 k\Omega$$

With this configuration the current through the voltage divider is 75  $\mu A$  at 8 V. The resolution of each RDCO step is then 200 mV.

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The value in RDCO register needed in order to have the desired output voltage can be calculated using the following formula:

$$RDCO = \frac{ldco}{2uA} = \frac{lR2 - lR1}{2uA} = \frac{\frac{0.5}{6.7k} - \frac{Vout - 0.5}{100k}}{2uA}$$

The resulting value will vary slightly depending on the tolerances of the selected resistors and DCO current.

The voltage is reduced from 8 V (RDCO = 0) to 2.5 V, when RDCO = 27.

# DC/DC step up operation: regulation at voltages higher than power supply

A practical application of the present case is the control of blue lasers. This type of laser present a forward voltage around 5 V, which demands an LDA voltage of about 6 V. If the system is supplied with a 3 V LiPo battery, it is necessary to use a the DC/DC in order to step up and drive the laser diode and driver with a sufficient voltage. Figure 21 shows this application:



#### Figure 21: Regulation at voltage greater than power supply

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Setting register RDCO to 10 it delivers 20  $\mu A$  and 6 V are obtained at Vout.

iC-HT must be supplied by a voltage within the thresh-

Extension of system working voltage range

old values of 2.8 and 8 V. It is possible to control the DC/DC output in a voltage range of 2.5 to 12 V if the DC/DC converter, controlled by DCO output signal, is included in the system, as it is shown in Figure 22:



Figure 22: Extension of working voltage range

In the Figure 22 both the laser and the iC-HT are supplied by output voltage Vout from DC/DC converter. The register RDCO is set to 23, which forces 48  $\mu$ A to be output to the voltage divider. A system voltage of 3.3 V is obtained at Vout.

#### Efficiency enhancement

If iC-HT and the laser diode are supplied with the same power supply, the efficiency of the driver can be poor,

depending on the supplied voltage, the saturation voltage and the laser diodes forward voltage. Power dissipation of the driver can be reduced if LDAx is fed through the DC/DC converter configured to deliver a lower voltage than the power supply as shown in Figure 23.

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Figure 23: System efficiency enhancement

The register RDCO is set to 23 to provide a laser diode supply voltage of 3.3 V at Vout.

#### WATCHDOG TIMER

The internal 200 kHz oscillator is monitored with the Watchdog Timer (WDT).

If the oscillator remains longer than the maximum time of tWDT (cf. *Electrical Characteristics No. E03*) without activity an oscillator error is triggered. An oscillator error sets OSCERR error bit to 1. The automatic offset compensation of the laser control requires the oscillator.

The state of OSCERR is signaled at pin NCHK. The signaling of OSCERR state can be masked with bit MOSCERR. Setting MOSCERR to 1 masks the oscillator error and OSCERR will not be signaled at NCHK.

It is possible to simulate an error of the oscillator using SOSCERR bit. If SOSCERR = 1, the oscillator error is forced. When OSCERR is set to 1 the error will be signaled through NCHK, depending on the state of MOSCERR.

OSCERR	Addr. 0x00; bit 6	R
Code	Function	
0	Oscillator operates OK	
1	Watchdog timeout set on oscillator failure	

Table 57: Oscillator error

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MOSCERR	Addr. 0x1D; bit 0 R/W 0		S
Code	Function		С
0	Oscillator error (watchdog) will be signaled at NCHK	1	0
1	Oscillator error (watchdog) will not be signaled at NCHK		1
		,	

SOSCERR	Addr. 0x1D; bit	7 R/W	/ 0
Code	Function		
)	No oscillator error simul	ated	
1	Oscillator error simulate	d (watchdog timeout)	
_			

Table 58: Oscillator error mask

#### Table 59: Simulate oscillator error

#### TEMPERATURE MONITOR AND PROTECTION

iC-HT includes an 8 bit temperature monitor that allows to measure the internal chip temperature going from -40 to 125 °C. The resolution is 1 °C/LSB. The internal temperature can be obtained by reading register TEMP. TEMP is a read-only register.

TEMP	Addr. 0x02; bit 7:0	R
Code	Function	
0x00	Minimum temperature	
0xFF	Maximum temperature	

Table 60: Chip temperature

Absolute read values may differ from one chip to another. An individual initial calibration of the temperature monitor is recommended. The TEMP register must be read at a known temperature. Using the resolution value of 1 °C/LSB, the internal temperature can be calculated.

The temperature monitor can be used to compensate temperature effects on the laser diode. The microcontroller can use a laser diode characteristic formula or a look-up table combined with the temperature value measured through register TEMP. The reference voltage can be configured accordingly in order to compensate temperature effects.

iC-HT is protected against overtemperature. In iC-WK mode, if the internal temperature value exceeds the overtemperature threshold an OVT error event will be triggered and signaled through pin NCHK. Both laser channels will be disabled. Pin NCHK will keep signaling the error although the internal temperature goes down to a safe value below the overtemperature threshold value. If the temperature has exceeded the overtemperature threshold value, pins EC1 and EC2 have to be

pulled lo in order to stop signaling the error. Setting pin ECx back hi will re-enable the corresponding channel.

In microcontroller mode, if the internal temperature exceeds a safety value an overtemperature error bit (OVT) will be set to 1. If OVT = 1, both channels will be disabled and the error event will be signaled through pin NCHK. If the internal temperature goes down to a safe value below the overtemperature threshold value, OVT will remain at value 1. Reading the OVT bit stop signaling error through pin NCHK. Reading OVT bit will set it back to 0. Setting pin ECx Io and then back hi will allow re-enabling the corresponding channel.

The overtemperature threshold value can not be configured.

OVT	Addr. 0x00; bit 3	R
Code	Function	
0	No overtemperature event has occurred since last read	
1	Overtemperature event has occurred. Cleared on read	

Table 61: Overtemperature detection

In microcontroller mode it is possible to simulate an overtemperature event using bit SOVT. Setting SOVT to 1, the overtemperature error flag OVT will be set to 1. iC-HT will remain in error state until SOVT is set back to 0.

SOVT	Addr. 0x1D; bit 4	R/W 0
Code	Function	
0	No overtemperature event is simulated.	
1	Overtemperature event simulated.	

Table 62: Simulate overtemperature

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#### DIGITAL INTERFACE AND MEMORY INTEGRITY MONITOR

iC-HT provides a microcontroller slave interface by selection on the EMC pin. iC-HT support the interfaces SPI or I<sup>2</sup>C that are selected by the INS/WKR pin.

registers) to be made prior to apply it to the laser channels. iC-HT has two different modes selectable by the MODE(1:0) register (address 0x1C).

EMC	C Addr. Pin;	
Code	Code Function	
lo	lo iC-WK-mode, digital interfaces disabled	
Open	Not allowed, error signaled	
hi	MCU mode, interface selected by INS/WKR enabled	

Table 63: Enable microcontroller

INS/WKR	Addr. Pin;	
Code	Function	
lo	SPI interface selected	
Open	Not allowed, error signaled.	
hi	I <sup>2</sup> C interface selected	

Table 64: Interface selection I<sup>2</sup>C or SPI

The configuration memory is integrity monitored and atomic executable (all at once: changes of the configurations without any direct effects, the changes are executed at once by command ) to the functional blocks of iC-HT.

Integrity monitoring is implemented by a duplication of the configuration registers into a validation page (see description below) where the register are automatically copied with inverted value. Every register bit is compared with its validation copy and in case of difference, a memory error is generated and both laser channels are switched off.

Atomic appliance is achieved by latching the configuration registers. This permits a full configuration (different

MODE(1:0)	Addr. 0x1C; bit 1:0	R/W 01
Code	Function	
00	Invalid parameter	
01	Operation mode	
10	Configuration mode	
11	Invalid parameter	

Table 65: Configuration and operation mode

In Configuration mode, the configuration memory (address 0x10 to 0x1F) can be written and read back to check a correct communication without changing the present configured operation state of the iC-HT. In this mode, the memory integrity check is disabled.

iC-HT will monitor the time elapsed in configuration mode and automatically switch the laser off if it exceeds a configuration mode timeout. The time in configuration mode must less than 40 ms for ensuring that no configuration timeout occurs during configuration (cf. Electrical Characteristics No. E02). The timeout can be up to 164 ms.

When writing the configuration is completed, iC-HT is switched to operation mode by writing "0b01" into the MODE register (address 0x1C). In operation mode the configuration is applied to the iC-HT and the memory integrity check activated. In this mode configuration registers can only be read (except MODE(1:0) register, which is always write accessible). Figure 24 shows the interface to memory structure.



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Figure 24: Interface, RAM integrity monitoring and configuration latching

#### **Register map description**

The register map consists of 64 addresses subdivided in three different pages:

- Read-only page, address 0x00 to 0x0F: iC-HT status, ADC readouts, thermometer readout and chip revision.
- Configuration page (integrity monitored), readwrite registers, address 0x10 to 0x1F.
- Validation page, read-write registers, address 0x30 to 0x3F.

#### Read-only registers

Read-only registers are sub-divided as well into status registers (address 0x00 to 0x01) and measurement registers. Status registers are normally latched to 1 on events and cleared on read (see individual register description). Measurement registers are dual-port and can be accessed simultaneously with the measurements in progress. ADC1(address 0x03 to 0x04) and ADC2 (address 0x05 to 0x06) are 10 bit registers split into two 8 bit registers each and must be accessed in block mode (automatic address increment) to ensure data not changing during the read.

#### Configuration page (integrity monitored)

The configuration page (address 0x10 to 0x1F) contains the registers that control the driver. Every write operation to any of the registers of this page will be internally duplicated to the correspondent register at the validation page. After the write operation, the correspondent validation register will contain the inverted value of the configuration register.

#### Validation page

The validation page (address 0x30 to 0x3F) can be read or written normally. Only when a write procedure is made to any of the configuration registers the correspondent validation pair will be written with the inverted value of the configuration register as well.

Both the configuration and validation pages are initialized during power-up. This event is signaled at the STATUS0 register (bit 0, INITRAM). In standby mode (NSTBY = Io) the RAM is not reset if any write command has been executed and therefore, configuration and validation pages keep the stored information and INITRAM remains unset. Entering standby mode after power-up without any write command, the RAM will be initialized again and the INITRAM bit will be set to 1 again. Any VDD power-down event signaled at the STA-TUS0 register outside the standby mode (NSTBY = hi) requires a RAM content check regardless of the state of the INITRAM bit to ensure data is not corrupted.

#### Possible start-up sequence:

- iC-HT starts in operation mode with default configuration. INITRAM and PDOVDD error bits must be set in STATUS0, DISC1 (address 0x10, bit 3) and DISC2 (address 0x15, bit 3) are set to 1.
- 2. Write MODE(1:0) = "10" register (Addr. 0x1C) to enable the configuration mode.
- 3. Configure the laser channels.
- 4. Read back to verify a correct data transfer.
- 5. Set the DISC1, DISC2 bits to 0 on used channels.
- Read the status registers(address 0x00, 0x01, 0x02) to detect possible errors and validate status. If any error exist, read again to ensure its validation.
- Write MODE(1:0) = "01" register (address 0x1C) to apply the configuration and enable the memory integrity check.
- 8. During operation: monitor the status registers checking for errors. The NCHK pin signals any set status bit if not masked. This pin can be used to trigger an microcontroller interrupt line.

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# **iC-HT** DUAL CW LASER DIODE DRIVER START-UP Setting pin NSTBY to lo iC-HT enters standby mode. In stand by mode and with no supply voltage at pin VDD time (Ten), EC1 be switched on

After wake-up (pin NSTBY rising edge), the internal regulated supply VDD is generated again. The required time Tvdd depends on the capacitor connected to pin VDD (cf. *Electrical Characteristics No. 504*).

and the current consumption on VB is reduced to less

than 10 µA (cf. Electrical Characteristics No. 002).

Once the VDD voltage level is correct, iC-HT enters an offset compensation procedure regardless of the state of the laser enable pins (EC1, EC2). During this time (Ten), EC1 and EC2 are ignored and laser cannot be switched on (cf. *Electrical Characteristics No. 111*). After this time (Ten), laser channels can be switched on.

The switch-on procedure needs an initial time (Tci) to reach the 80% of the target light power (in APC mode) or laser current (in ACC mode) (cf. *Electrical Characteristics No. 112*) and an additional time (Tcio) to reach the 99% of the value (cf. *Electrical Characteristics No. 113*). Figure 25 illustrates an startup example for channel 1 in iC-WK mode.



605/24 / V5 / MaH-IF / ichaus/ht-dual-cw-laser-diode-driver

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#### **REGISTER OVERVIEW**

OVERV	'IEW								
Addr	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	
0x00 R	CFGTIMO	OSCERR	OVC1	OVC2	OVT	MEMERR	PDOVDD	INITRAM	
0x01 R		LDKSAT2	MONC2	MAPC2		LDKSAT1	MONC1	MAPC1	
0x02 R	TEMP(7:0)								
0x03 R							ADC	21(9:8)	
0x04 R				ADC	1(7:0)	•			
0x05 R							ADC	2(9:8)	
0x06 R		•		ADC	2(7:0)				
0x07 R				Not imp	emented				
	Not implemented								
0x0F R	Chip revision mark								
0x10		ADCC1(2:0)		EOC1	DISC1	DISP1	ECIE1	EACC1	
0x11	ILIM1(7:0)								
0x12	RMD1(7:0)								
0x13		COMP1(2:0) RLDKS1(1:0)					REF1(9:8)		
0x14	REF1(7:0)								
0x15		ADCC2(2:0)		EOC2	DISC2	DISP2	ECIE2	EACC2	
0x16				ILIM	2(7:0)				
0x17				RMD	2(7:0)				
0x18			COMP2(2:0)		RLDK	S2(1:0)	REF	2(9:8)	
0x19				REF	2(7:0)				
0x1A		ADSNF2		RACC2		ADSNF1		RACC1	
0x1B		MERGE			RDC	O(5:0)			
0x1C			Not impl	emented			MOE	DE(1:0)	
0x1D	SOSCERR	SOVC2	SOVC1	SOVT	MLDKSAT2	MLDKSAT1	MMONC	MOSCERF	
0x1E				Reserved regis	ster. Set to zero				
0x1F			Rese	rved register(Fa	ctory test). Set t	o zero			
0x20				Not imp	emented				
				Not imp	emented				
0x30			V	alidation content	t for 0x10, invert	ed			
0x31			V	alidation conten	t for 0x11, inverte	ed			
0x3F			V	alidation conten	t for 0x1F, invert	ed			

Table 66: Register layout

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#### PARAMETERS

Register	Address	Bits	Default	Description
ADCC1	0x10	7:5	000	Configuration for ADC from channel 1
ADCC2	0x15	7:5	000	Configuration for ADC from channel 2
ADSNF1	0x1A	2	0	MDA force/sense for ADC measurement in channel 1
ADSNF2	0x1A	6	0	MDA force/sense for ADC measurement in channel 2
RACC1	0x1A	0	0	Channel 1 ACC resistor mirror factor
RACC2	0x1A	4	0	Channel 2 ACC resistor mirror factor
COMP1	0x13	6:4	011	Channel 1 regulator delay compensation
COMP2	0x18	6:4	011	Channel 2 regulator delay compensation
DISC1	0x10	3	1	Software disable for channel 1
DISC2	0x15	3	1	Software disable for channel 2
DISP1	0x10	2	0	Disable PLR for channel 1
DISP2	0x15	2	0	Disable PLR for channel 2
EACC1	0x10	0	0	Enable ACC mode for channel 1
EACC2	0x15	0	0	Enable ACC mode for channel 2
ECIE1	0x10	1	0	Enable external CI capacitor for channel 1
ECIE2	0x15	1	0	Enable external CI capacitor for channel 2
EOC1	0x10	4	1	Enable offset compensation for channel 1
EOC2	0x15	4	1	Enable offset compensation for channel 2
ILIM1	0x11	7:0	0xFF	Current limit at channel 1
ILIM2	0x16	7:0	0xFF	Current limit at channel 2
MERGE	0x1B	6	0	MERGE channels 1 and 2, controlled by channel 1
MLDKSAT1	0x1D	2	1	LDKSAT1 error mask
MLDKSAT2	0x1D	3	1	LDKSAT2 error mask
MMONC	0x1D	1	1	MONC error mask
MODE	0x1C	1:0	01	Configuration / Operation mode selection
MOSCERR	0x1D	0	0	OSCERR error mask
RDCO	0x1B	5:0	0x02	DC converter set point
REF1	0x13/0x14	9:0	0x000	Voltage reference at channel 1
REF2	0x18/0x19	9:0	0x000	Voltage reference at channel 2
RLDKS1	0x13	3:2	00	Channel 1 LDK saturation detector threshold
RLDKS2	0x18	3:2	00	Channel 2 LDK saturation detector threshold
RMD1	0x12	7:0	0xFF	Resistor at channel 1
RMD2	0x17	7:0	0xFF	Resistor at channel 2
SOSCERR	0x1D	7	0	Oscillator error simulation (watchdog timeout)
SOVC1	0x1D	5	0	Overcurrent event at channel 1 simulation
SOVC2	0x1D	6	0	Overcurrent event at channel 2 simulation
SOVT	0x1D	4	0	Overtemperature event simulation
Reserved	0x1A	7:0	0x00	Reserved
Reserved	0x1F	7:0	0x00	Reserved

Table 67: Parameter overview

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R

Register	Address	Bits	Default	Description
INITRAM	0x00	0	R/O	RAM initialized.
PDOVDD	0x00	1	R/O	Power-down event at VDD
MEMERR	0x00	2	R/O	RAM memory validation error
OVT	0x00	3	R/O	Overtemperature event
OVC2	0x00	4	R/O	Overcurrent at channel 2
OVC1	0x00	5	R/O	Overcurrent at channel 1
OSCERR	0x00	6	R/O	Oscillator error (watchdog set)
CFGTIMO	0x00	7	R/O	Configuration mode timeout event
MAPC1	0x01	0	R/O	Channel 1 current state
MONC1	0x01	1	R/O	Monitor channel 1 enabled at least once (latched)
LDKSAT1	0x01	2	R/O	Channel 1 LDK saturation event
MAPC2	0x01	4	R/O	Channel 2 current state
MONC2	0x01	5	R/O	Monitor channel 2 enabled at least once (latched)
LDKSAT2	0x01	6	R/O	Channel 2 LDK saturation event

Table 68: Status overview

Register	Address	Bits	Default	Description
TEMP	0x02	7:0	R/O	Chip temperature measurement
ADC1h	0x03	1:0	R/O	Channel 1 ADC 9:8 readout
ADC1I	0x04	7:0	R/O	Channel 1 ADC 7:0 readout
ADC2h	0x05	1:0	R/O	Channel 2 ADC 9:8 readout
ADC2I	0x06	7:0	R/O	Channel 2 ADC 7:0 readout
CHIPREV	0x0F	7:0	R/O	Chip revision identification

#### Table 69: Measurement overview

PDOVDD

#### **Device identification**

Addr. 0x0F; bit 7:0	R
Function	
Initial version iC-HT	
iC-HT_1	
iC-HT_Z	
Reserved	
	Function Initial version iC-HT iC-HT_1 iC-HT_Z

Table 70: Device identification

Code	Function					
0	VDD power down not occurred since last read					
1	VDD power down event has occurred. Cleared on read					
	Table 72: VDD power down					
MEMERR	Addr. 0x00; bit 2 R					
Code	Function					
0	RAM has not been changed since last validation	_				

Addr. 0x00; bit 1

## Status

Table 73: Memory validation

RAM has changed and has not been validated

			OVT	Addr. 0x00; bit 3	R
INITRAM	Addr. 0x00; bit 0	R	Code	Function	
NIIKAW	Addr. 0x00, bit 0	ĸ	0	No overtemperature event has occurred	d since last
Code	Function		ľ	read	
)	RAM not initialized since last read		1	Overtemperature event has occurred. C	cleared on
1	RAM initialized. Cleared on read			read	
	Table 71: RAM initialization			Table 74: Overtemperature	

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Table 71: RAM initialization

Table 74: Overtemperature

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OVC2	Addr. 0x00; bit 4	R
Code	Function	
0	No overcurrent event at channel 2 has occurred since last read	
1	Overcurrent event at channel 2 has occurred. Cleared on read	

Table 75: Overcurrent channel 2

No overcurrent event at channel 1 has occurred

Overcurrent event at channel 1 has occurred.

Table 76: Overcurrent channel 1

Watchdog timeout set on oscillator failure.

iC-HT not in Configuration Mode or Timeout did not

iC-HT in Configuration Mode and Timeout happened. Laser switched off.

Channel 1 is off at the precise reading moment Channel 1 is on at the precise reading moment

Table 77: Oscillator watchdog

Table 78: Configuration timeout

Addr. 0x01; bit 0

Table 79: EC1 pin state

Addr. 0x01; bit 1

Addr. 0x00; bit 7

Addr. 0x00; bit 6

Oscillator functioning OK

R

R

R

R

R

Addr. 0x00; bit 5

Function

Function

Function

Function

Function

Cleared on read

happened till now

since last read

Cleared on read

OVC1

Code

OSCERR

CFGTIMO

Code

MAPC1

MONC1

Code

0

1

Code

0

1

0

1

Code

0

1

0

1

LDKSAT1	Addr. 0x01; bit 2	2
Code	Function	
0	Channel 1 LDK saturation voltage not reached.	
1	Channel 1 LDK saturation voltage reached at least once, cleared on read	t

Table 81: LDK1 saturation

MAPC2	Addr. 0x01; bit 4	R
Code	Function	
0	Channel 1 is off at the precise reading moment	
1	Channel 1 is on at the precise reading moment	

#### Table 82: EC2 pin state

MONC2	Addr. 0x01; bit 5	R
Code	Function	
0	Channel 2 has not been switched on since	
1	Channel 2 has not been switched on since	last read

Table 83: Monitor channel 2

LDKSAT2	Addr. 0x01; bit 6 R	
Code	Function	
0	Channel 2 LDK saturation voltage not reached.	
1	Channel 2 LDK saturation voltage reached at least once, Cleared on read	

Table 84: LDK2 saturation

#### Measurement registers

TEMP	Addr. 0x02; bit 7:0	R
Code	Function	
0x00	Minimum temperature	
0xFF	Maximum temperature	

#### Table 85: Chip temperature

ADC1	Addr. 0x03/04; bit 9:0	R
Code	Function	
0x000	ADC minimum value	
0x3FF	ADC maximum value	

#### Table 86: ADC channel 1

ADC2	Addr. 0x05/06; bit 9:0	R
Code	Function	
0x000	ADC minimum value	
0x3FF	ADC maximum value	

Table 87: ADC channel 2

Table 80: Monitor channel 1

Channel 1 has not been switched on since last read Channel 1 has been switched on at least once.

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channel 1



#### **Channel 1 configuration registers**

Function

EACC1

Code

0

1

ILIM1	Addr. 0x11; bit 7:0	R/W 0xFF
Code	Function	
0x0A	Channel 1 overcurrent threshold s current	et to the minimum
	Channel 1 overcurrent threshold s $llim = (\Delta l(LDK) \cdot n)$ , n from 10 to	
0xFF	Channel 1 overcurrent threshold se current	et to the maximum

Table 94: Overcurrent threshold configuration channel 1

RMD1	Addr. 0x12; bit 7:0	R/W 0xFF
Code	Function	
0x00	PLR1 set to the minimum resistanc	e
	PLR1 resistor set to Rmd = Rmd <sub>0</sub> ( n from 0 to 255	$1 + \frac{\Delta Rmd(\%)}{100})^{n+1}$ ,
0xFF	PLR1 resistor set to the maximum r	resistance

Table 95: MDA resistor channel 1

COMP1	Addr. 0x13; bit 6:4	R/W 011
Code	Function	
000	Minimum regulator delay compensation for channel 1, slower response	
111	Maximum regulator delay compensation	
	for channel 1, faster response	

Table 96: Regulator delay compensation channel 1

RLDKS1	Addr. 0x13; bit 3:2	R/W 00
Code	Function	
00	V(LDK1) < 0.5 V sets the LDKSAT1 alarm	bit
01	V(LDK1) < 0.8 V sets the LDKSAT1 alarm	bit
10	V(LDK1) < 1.0 V sets the LDKSAT1 alarm	bit
11	V(LDK1) < 1.2 V sets the LDKSAT1 alarm	bit

#### Table 97: LDK saturation threshold selection channel 1

REF1	Addr. 0x13/14; bit 9:0 R/W 0x000
Code	Function
0x000	Channel 1 regulator reference voltage set to minimum voltage
	Channel 1 regulator reference voltage set to $Vref = Vref_0(1 + \frac{\Delta Vref(\%)}{100})^{n+1}$ , n from 0 to 1023
0x3FF	Channel 1 regulator reference voltage set to maximum voltage

#### Table 98: Regulator voltage reference channel 1

RACC1	Addr. 0x1A; bit 0	R/W 0
Code	Function	
0	Current range high for channel 1	
1	Current range low for channel 1	

Table 99: Current range configuration channel 1

regulation) ACC mode enabled for channel 1 (laser current regulation)

APC mode enabled for channel 1 (light power

Addr. 0x10; bit 0

R/W 0

Table 88: Enable APC/ACC channel 1

ECIE1	Addr. 0x10; bit 1 R	/W 0
Code	Function	
0	External CI capacitor for channel 1 disconnected	
1	External CI capacitor for channel 1 connected	

Table 89: Enable external CI capacitor channel 1

DISP1	Addr. 0x10; bit 2	R/W 0
Code	Function	
0	PLR enabled for channel 1	
1	PLR disabled for channel 1	

Table 90: Disable PLR channel 1

DISC1	Addr. 0x10; bit 3	R/W 1
Code	Function	
0	Channel 1 can be enabled by EC1 pin	
1	Channel 1 cannot be enabled by EC1 pin	

Table 91: Disable channel 1

EOC1	Addr. 0x10; bit 4	R/W 1
Code	Function	
0	Channel 1 regulator offset compensation disabled	
1	Channel 1 regulator offset compensation enabled	

Table 92: Enable offset compensation channel 1

ADCC1(2:0)	Addr. 0x10;	bit 7:5	R/W 000
Code	Function		
0xx	Channel 1 ADC disa	abled	
100	Channel 1 ADC sou	rced by V(MD/	A1), ADSNF1 = 0
100	Channel 1 ADC sou	rced by V(PLR	(1), ADSNF1 = 1
101	Channel 1 ADC sou	rced by V(VB)	
110	Channel 1 ADC sou	rced by V(VDE	0)
111	Channel 1 ADC sou	rced by V(LDK	(1)

#### Table 93: ADC source selection channel 1

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R/W 0xFF

ADSNF1	Addr. 0x1A; bit 2	R/W 0
Code	Function	
0	ADC measurement MDA1 pad (force)	
1	ADC measurement PLR1 (sense)	

Table 100: ADC channel 1 sense/force selection

#### ADCC2(2:0) Addr. 0x15; bit 7:5 R/W 000 Function Code 0xx Channel 2 ADC disabled Channel 2 ADC sourced by V(MDA2), ADSNF2 = 0 100 100 Channel 2 ADC sourced by V(PLR2), ADSNF2 = 1 101 Channel 2 ADC sourced by V(VB) 110 Channel 2 ADC sourced by V(VDD) 111 Channel 2 ADC sourced by V(LDK2)

**Channel 2 configuration registers** 

Table 106: ADC source selection channel 2

Addr. 0x16; bit 7:0

ILIM2

EACC2	Addr. 0x15; bit 0	R/W 0
Code	Function	
0	APC mode enabled for channel 2 (light power regulation)	
1	ACC mode enabled for channel 2 (laser current regulation)	

Table 101: Enable APC/ACC channel 2

Code	Function
0x0A	Channel 2 overcurrent threshold set to the minimum current
	Channel 2 overcurrent threshold set to $llim = (\Delta l(LDK) \cdot n)$ , n from 10 to 255
0xFF	Channel 2 overcurrent threshold set to the maximum current

Table 107: Overcurrent threshold configuration channel 2

ECIE2	Addr. 0x15; bit 1	R/W 0
Code	Function	
0	External CI capacitor for channel 2 disconnected	
1	External CI capacitor for channel 2 connected	

 Table 102: Enable external CI capacitor channel 2

DISP2	Addr. 0x15; bit 2	R/W 0
Code	Function	
0	Internal resistor at MDA2 enabled for channel 2	
1	Internal resistor at MDA2 disabled for channel 2	

Table 103: Disable PLR channel 2

DISC2	Addr. 0x15; bit 3	R/W 1
Code	Function	
0	Channel 2 can be enabled by EC2 pin	
1	Channel 2 cannot be enabled by EC2 pin	

Table 104: Disable channel 2

EOC2	Addr. 0x15; bit 4	R/W 1
Code	Function	
0	Channel 2 regulator offset compensation disabled	
1	Channel 2 regulator offset compensation enabled	

Table 105: Enable offset compensation channel 2



Table 108: MDA resistor channel 2

COMP2	Addr. 0x18; bit 6:4	4 R/W 011
Code	Function	
000	Minimum regulator delay of for channel 2, slower response	
 111	Maximum regulator delay for channel 2, faster respo	

Table 109: Regulator delay compensation channel 2

RLDKS2	Addr. 0x18; bit 3:2	R/W 00
Code	Function	
00	V(LDK2) < 0.5 V sets the LDKSAT2 ala	rm bit
01	V(LDK2) < 0.8 V sets the LDKSAT2 alarm bit	
10	V(LDK2) < 1.0 V sets the LDKSAT2 alarm bit	
11	V(LDK2) < 1.2 V sets the LDKSAT2 alarm bit	

Table 110: LDK saturation threshold selection channel 2

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R/W 0

R/W 1

REF2	Addr. 0x18/19; bit 9:0	R/W 0x000
Code	Function	
0x000	Channel 2 regulator reference vol minimum voltage	tage set to
	Channel 2 regulator reference voltage set to $Vref = Vref_0(1 + \frac{\Delta Vref(\%)}{100})^{n+1}$ , n from 0 to 1023	
0x3FF	Channel 2 regulator reference vol maximum voltage	tage set to

Code	Function	
0	Current range high for channel 2	
1	Current range low for channel 2	
Table 112: Current range configuration channel 2		

Addr. 0x1A; bit 4

RACC2

MMONC

Code

0

1

1

Table 111: Regulator voltage reference channel 2

ADSNF2	Addr. 0x1A; bit 6	R/W 0
Code	Function	
0	ADC measurement MDA2 pad (force)	
1	ADC measurement PLR2 (sense)	

Table 113: ADC channel 2 sense/force selection

Addr. 0x1D; bit 1

MONC1 and MONC2 event will be signaled at NCHK

MONC1 and MONC2 event

will not be signaled at NCHK

Function

#### General configuration registers

RDCO	Addr. 0x1B; bit 5:0	R/W 0x02
Code	Function	
0x00	No current	
0x3F	140 µA Typ (see spec point D01)	

Table 114: DCO current control

Table 118: Monitor channel 1 and 2 event mask

MERGE	Addr. 0x1B; bit 6	R/W 0
Code	Function	
0	Channel 1 and 2 operate independently	/
1	Power transistor from channel 2 usable in parallel with channel 1, regulation made by channel 1.	

Table 115: Channel combination

MODE(1:0)	Addr. 0x1C; bit 1:0	R/W 01
Code	Function	
00	Not allowed, signaled as memory error	
01	Chip set in operation mode (apply configuration, latch transparent)	
10	Chip set in configuration mode (hold previous configuration)	
11	Not allowed, signaled as memory error	

Table 116: Configuration and operation mode

MOSCERR	Addr. 0x1D;	bit 0	R/W 0
Code	Function		
0	Oscillator error (watchdog) will be signaled at NCHK		
1	Oscillator error (watchdog) will not be signaled at NCHK		

Table 117: Oscillator watchdog error mask

MLDKSAT1 R/W 1 Addr. 0x1D; bit 2 Code Function 0 LDKSAT1 event will be signaled at NCHK LDKSAT1 event will not be signaled at NCHK

Table 119: LDK saturation error mask channel 1

MLDKSAT2	Addr. 0x1D; bit 3	R/W 1
Code	Function	
0	LDKSAT2 event will be signaled at NCHK	
1	LDKSAT2 event will not be signaled at NC	нк
	Code	Code Function

Table 120: LDK saturation error mask channel 2

SOVT	Addr. 0x1D; bit 4	R/W 0
Code	Function	
0	No overtemperature event is simulated.	
1	Overtemperature event simulated.	

#### Table 121: Simulate overtemperature

SOVC1	Addr. 0x1D; bit 5	R/W 0
Code	Function	
0	No Overcurrent event at channel 1 is simulated.	
1	Overcurrent event at channel 1 simulated.	
0 1		

Table 122: Simulate overcurrent channel 1

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SOVC2	Addr. 0x1D; bit 6	R/W 0	SOSCERR	Addr. 0x1D;
Code	Function		Code	Function
0	No overcurrent event at channel 2 is	simulated.	0	No oscillator error s
1	Overcurrent event at channel 2 simulated.		1	Oscillator error simu

Table 123: Simulate overcurrent channel 2

Table 124: Simulate oscillator error

#### **DESIGN REVIEW: Notes On Chip Functions**

iC-HT 0, 1				
No.	Function, Parameter/Code	Description and Application Notes		
1	Adr. 0x1A	Reserved, Adr. 0x1A Should be always zero		
2	CHIPREV	For iC-HT chip releases see Table 70.		

Table 125: Notes on chip functions regarding iC-HT chip release 0 and 1.

iC-HT Z		
No.	Function, Parameter/Code	Description and Application Notes
1	ADSNF1,ADSNF2	Adr. 0x1A bits 2, 6 new implemented. See page nr. 28
2	RACC1,RACC2	Adr. 0x1A bits 0, 4 new implemented. See page nr. 17
3	CHIPREV	For iC-HT chip releases see Table 70.

Table 126: Notes on chip functions regarding iC-HT chip release Z

#### **REVISION HISTORY**

Rel.	Rel. Date*	Chapter	Modification	Page
A1	13-02-22		Initial Release.	
Rel.	Rel. Date*	Chapter	Modification	Page
B1	15-02-09	ELECTRICAL CHARACTERISTICS	Item 001: VB max updated from 8 V to 11 V.	6
		ELECTRICAL CHARACTERISTICS	Item 107: Laser overcurrent shutdown threshold updated from 93mA to 80mA	6
		ELECTRICAL CHARACTERISTICS	Item 108: Shutdown threshold resolution updated from 3 mA to 5 mA Item 108: Shutdown threshold resolution updated from 0.3 mA to 0.5 mA	6
		ELECTRICAL CHARACTERISTICS	Item 114: LDKx ACC mode current updated from 60 mA to 50 mA Item 114: LDKx ACC mode current updated from 750 mA to 650 mA Item 114: LDKx ACC mode current updated typical from 905 mA to 750 mA	7
		ELECTRICAL CHARACTERISTICS	Item 115: added TK -1500 ppm/K min, -500 ppm/K typ., 0 ppm/K max	7
		ELECTRICAL CHARACTERISTICS	Item 302: Percentual voltage increments updated from 0.05% to 0.1%	7
		ELECTRICAL CHARACTERISTICS	Item 403: Leakage Current at NCHK updated from ±10 to ±1	7
		ELECTRICAL CHARACTERISTICS	Item 602: Input Threshold Voltage at 2.8 V updated from 0.7 V to 0.6 V	7
		ELECTRICAL CHARACTERISTICS	Item B02: Overtemperature Release updated from 160 °C to 165 °C	8
		OPERATING AND STANDBY MODES	EMC and INS open pin description updated Web link to iC-WK iC-WKL updated	10
		MICROCONTROLLER MODE	MAPCx description updated Figure 10 updated with more details	20
		COMBINING BOTH CHANNELS	ILIM1 and REF1 tables added	22
		SERIAL COMMUNICATION INTERFACES	Figure 14: I <sup>2</sup> C frames updated	24

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PARAMETERS	Table 81: MAPC1 description updated Table 84: MAPC2 description updated	40
PARAMETERS	Table 114: RACC2 updated	42
PARAMETERS	Table 120: MMONC description updated	43

Rel.	Rel. Date*	Chapter	Modification	Page
C1		BLOCK DIAGRAM	Updated	1
		ELECTRICAL CHARACTERISTICS	Items 004, 005, 007, 008 updated	6
		ELECTRICAL CHARACTERISTICS	Items 601, 602, 603: Parameters updated	7
		iC-WK MODE	Figure 6 new	13
		MICROCONTROLLER MODE	Description updated	16
		MICROCONTROLLER MODE	Figure 11 updated	21
		MICROCONTROLLER MODE	Figure 12 new	21
		SERIAL COMMUNICATION INTERFACES	Figure 16 updated	24
		8 BIT INTERNAL PROGRAMMABLE LOGARITHMIC MONITOR RESISTORS	Figure 17 new	26
		10 BIT LINEAR A/D CONVERTER	Description updated	28
		WATCHDOG TIMER	Table 57 updated	32
		DIGITAL INTERFACE AND MEMORY INTEGRITY MONITOR	Description updated	34
		PARAMETERS	Table 77 updated	40

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\* Release Date format: YYYY-MM-DD

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#### ORDERING INFORMATION

Туре	Package	Order Designation
iC-HT	QFN28 5 mm x 5 mm	iC-HT QFN28-5x5
Evaluation Board	100 mm x 80 mm eval board	iC-HT EVAL HT1D

Please send your purchase orders to our order handling team.

For technical support, information about prices and terms of delivery please contact us.

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